

Variation-Aware Deterministic ATPG

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Abstract—In technologies affected by variability, the detection status of a small-delay fault may vary among manufactured circuit instances. The same fault may be detected, missed or provably undetectable in different circuit instances. We introduce the first complete flow to accurately evaluate and systematically maximize the test quality under variability. As the number of possible circuit instances is infinite, we employ statistical analysis to obtain a test set that achieves a fault-efficiency target with an user-defined confidence level. The algorithm combines a classical path-oriented test-generation procedure with a novel waveform-accurate engine that can formally prove that a small-delay fault is not detectable and does not count towards fault efficiency. Extensive simulation results demonstrate the performance of the generated test sets for industrial circuits affected by uncorrelated and correlated variations.

Keywords—Variation-aware test, fault efficiency, ATPG

I. INTRODUCTION

Recent semiconductor manufacturing technology nodes are heavily affected by massive parameter variations which create severe challenges for design and test of state-of-the-art circuits [1, 2]. Manufacturing defects interact with the effects of variations in complex and often unpredictable ways. This holds in particular for *small-delay faults*, which are defined as fixed amounts of extra delay at a given gate. Traditional fault-coverage metrics and test-generation methods are insufficient for circuit populations affected by variations. They ignore the fact that a given test set could detect a fault in some manufactured instances of the circuit but miss the same fault in other instances. The fault may also be provably undetectable in some of the instances, and such faults should not count towards fault efficiency. However, the parameters affected by variability, i.e., delays, are continuous and the number of possible circuit instances is infinite.

A further substantial difficulty in variation-aware testing is the reliance of traditional test methods on the concept of *path sensitization* [3, 4]. It is assumed that, in order to detect a small-delay fault f on a given gate, it is sufficient to find a test pair that sensitizes the longest path that goes through that gate. However, even in absence of variations, such procedures are not suited to identify undetectable faults. They enforce certain *sensitization conditions*, including robust, non-robust and functional sensitization. If no robustly sensitizable path of sufficient length (that is, cumulative delay) exists through the fault location, the fault is not proven undetectable, as detection through non-robustly or functionally sensitized path

could be possible. On the other hand, non-robust or functional sensitization is prone to invalidation, and faults for which such a path has been found may actually not be detected [5]. This situation is even worse under variations where the same path may have a length that is sufficient for detection in one instance but not sufficient in another.

In this paper, we present a statistical metric that generalizes the concept of fault efficiency (FE) to circuits under variations. We assume that variations are described by a known probability distribution, which may be correlated or uncorrelated. The metric consists of three parts: fault efficiency bound FE_{\min} , probability c and confidence γ . An individual circuit instance i has fixed delays, and the fault efficiency (FE) of a test (pair) set T with respect to an instance i is the number of faults detected by T divided by the number of detectable faults in i . Under variations, there are infinitely many instances i , and the same test set T may have different FE on different instances. T achieves fault efficiency FE_{\min} with probability c if the share of instances on which the fault efficiency of T is equal to or greater than FE_{\min} is at least c . In other words, if a large number N of random circuit instances is drawn according to the known probability distribution, at least $c \cdot N$ of them will have $FE \geq FE_{\min}$ and at most $(1 - c) \cdot N$ may have $FE < FE_{\min}$.

We introduce an automatic test generation algorithm that produces a test set that achieves FE_{\min} with probability c . This property of the test set holds with confidence level γ . The probability distribution and the parameters FE_{\min} , c and γ are inputs of the algorithm. Larger values of these parameters result in a larger test set. In contrast to previous variation-aware test generation approaches [6–12], the proposed algorithm relies on accurate fault efficiencies of the generated test sets. For this reason, the algorithm employs a very recently introduced SAT-based test-generation engine *WaveSAT* [5] that works with waveform precision and does not rely on path sensitization. *WaveSAT* is able to accurately classify a given fault, that is, to generate a test pair that detects it or to prove that the fault is undetectable and should be removed from fault efficiency calculation.

We applied our algorithm to mid-size industrial circuits and were able to generate test sets for large values (≥ 0.98) of FE_{\min} , c and γ . The approach is validated on uncorrelated and correlated probability distributions. The scalability is achieved by complementing *WaveSAT* by a fast timing-aware path-

sensitization procedure *PHAETON* [13, 14] and an efficient parallel delay fault simulator *fsim* running on general-purpose graphics processing units (GPGPUs) [15].

The remainder of the paper is organized as follows. Section II provides background on variation-aware test and the metrics used in this paper. It also introduces the statistical model of fault detection with a guaranteed confidence level. Section III starts with a brief review of procedures which generate test pairs for faults in circuit instances with fixed delays. The discussion of these procedures and their properties is essential as they are the basis of the adaptive confidence-guided algorithm, which is presented in the remainder of Section III. Experimental results are reported in Section IV. Section V concludes the paper.

II. VARIATION-AWARE TEST

A. Circuit model under variations

We assume that variability affects the delays of all gates of a circuit while leaving its topology unchanged. Therefore, gate delays are random variables described by a probability distribution. While uncorrelated variations have been predicted to be dominant in the nanoscale technologies [16–18], our approach is based on Monte-Carlo experiments and therefore works independent of the distribution. We will report experimental results for both uncorrelated and uncorrelated variations. In this paper, we assume circuits composed of primitive gates and a simplified timing model: each gate g_j has an input-to-output delay p_j . Each p_j is a random variable, and a circuit C with n gates is completely described by a *parameter configuration* $p = (p_1, \dots, p_n)$. This model can be easily extended by incorporating different rising and falling delays, pin-to-pin delays and pattern-dependent delays, resulting in more random variables per circuit. It is also possible to include travel times on wires and pulse filtering.

A circuit $i = C[p]$ where all gate delays have fixed values (p_1, \dots, p_n) is called a *circuit instance* or simply an *instance*. The *nominal instance* i_0 has all delays set to the expected values of their distributions. When circuits are manufactured, each produced circuit is described by an instance. Manufacturing N circuits corresponds to a *Monte-Carlo experiment* where the parameters (p_1, \dots, p_n) of each instance are drawn according to the distribution. It is possible to simulate this process by drawing sets of random parameters without actually manufacturing the circuit. This *Monte-Carlo simulation* yields a set of instances called *population*. Properties that are valid for the simulated population of sufficient size hold for different populations, including the actual manufactured circuits, with some probability called *confidence level*.

B. Fault detection under variations

We consider small-delay faults (SDFs) that are defined at logic gates in the circuit. An SDF (g, s) results in a slow-down of gate g by s time units. An SDF is *detected* by a test pair (v_1, v_2) if the circuit affected by this SDF has an incorrect value on at least one output at the *observation time* t_{obs} . We define t_{obs} as the time when the circuit's outputs are read out.

t_{obs} typically equals the clock cycle duration, and a (fault-free) circuit instance in which all possible transitions are finished before t_{obs} is called *timing-correct*. We assume that the value of t_{obs} is not affected by variations and is hence equal for all circuit instances in the population. Under normal circumstances, the designer will set t_{obs} such that the majority of (fault-free) circuit instances in a population are timing-correct and only a few instances have delays (p_1, \dots, p_n) that are so large that t_{obs} is exceeded. Such instances contribute to the parametric yield loss and are not considered here.

Detection of an SDF in two individual instances of the same circuit, i_1 and i_2 , is illustrated in the left and the right part of Figure 1, respectively. The SDF of size ε is located on gate g_1 . Gate delays are indicated by numbers, and the transitions on the individual lines are shown for the test pair $AB = 11/01$. In general, an SDF can only be detected if there is (at least one) input-to-output path that includes g , has a length that exceeds $(t_{obs} - s)$ and is *sensitized*, i.e., has transitions at the outputs of all gates. In Figure 1, test pair $AB = 11/01$ sensitizes path A - g_1 - g_3 - g_4 . For example, the small-delay fault of size $\varepsilon = 3.5$ would be detected in instance i_1 (transition on the output at time 10.5, after $t_{obs} = 8$) and not detected in instance i_2 (output transition at time 7.5, before t_{obs}). However, this fault is detectable in instance i_2 through path B - g_1 - g_2 - g_4 , sensitized by a different test pair $AB = 11/10$ and inducing the output transition at time 8.5.

A fault is undetectable if there is no path of sufficient length ($\geq t_{obs} - s$) through its location. We refer to this case by the term *structural undetectability*. In our example, the SDF on g_1 with size $\varepsilon = 2$ is detected in instance i_1 but structurally undetectable in instance i_2 . If a sufficiently long path does exist, it may or may not be possible to find a test pair that sensitizes it such as to induce the incorrect value at time t_{obs} . In summary, a given test set T partitions the set F of all considered SDFs in circuit instance i into three parts: $F = F_{det}(i, T) \cup F_{nd}(i) \cup F_{miss}(i, T)$, where $F_{det}(i, T)$, $F_{nd}(i)$ and $F_{miss}(i)$ are the sets of faults detected by T in instance i , faults that are provably undetectable in instance i , and faults that are detectable in i but have been missed by T , respectively. Note that $F_{nd}(i)$ depends on the instance (the same fault could be undetectable in some instances and detectable in others) but not on test set T . The *fault efficiency* of a test set T with respect to a fault set F in an instance i is

$$FE(F, T, i) = \frac{|F_{det}(i, T)|}{|F \setminus F_{nd}(i)|}. \quad (1)$$

It is important to point out the significance of the set $F_{nd}(i)$ of undetectable faults. When stuck-at faults are considered, most circuits have no or very few undetectable faults. In case of small-delay faults considered here, the detection depends on the relationship between the *slack* of the sensitized path (the difference between t_{obs} and its length) and the size s of the SDF. If many SDFs have a size that is close to the slack of a sensitizable path, they may be detectable in some circuit instances but undetectable in others. Fault coverage metrics employed in previous work [19] that did not incorporate

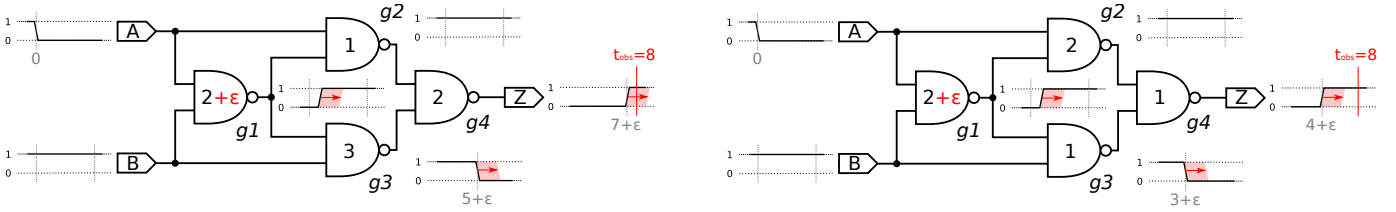


Fig. 1. Small-delay fault detection under variation.

accurate detectability status yielded quite low coverages (5% to 35%) for test sets that achieve 76% to 86% fault efficiency according to Eq. 1. Since our test generation is controlled by FE and we aim at rather high values (in excess of 98%), we employ a sophisticated undetectability check to each missed fault to decide whether it counts towards the denominator in Eq. 1 or not.

C. Statistical fault efficiency for circuit populations

The fault efficiency metric defined above is valid for one specific circuit instance i described by a given parameter configuration (p_1, \dots, p_n) . In this section, we introduce a statistical metric that holds for an *entire population* of circuit instances that share the same probability distribution of random variables (p_1, \dots, p_n) . As already mentioned in the introduction, this metric evaluates a test set T with respect to the three values FE_{min} , c and γ . For N randomly drawn circuit instances of a given parameter space, test set T achieves a fault efficiency of FE_{min} with probability c for an instance i (of the same distribution), if at least $\lceil (c \cdot N) \rceil$ out of the N instances have a fault efficiency that matches or exceeds FE_{min} :

$$\Pr[FE(F, T, i) \geq FE_{min}] \geq c. \quad (2)$$

Recall that calculating the fault efficiency (Eq. 1) requires classification of all considered faults into sets $F_{det}(i, T)$, $F_{nd}(i)$ and $F_{miss}(i)$.

Since the number of different instances in the parameter space is infinite, it is impossible to evaluate Eq. 2 directly. Instead, FE_{min} will be estimated with respect to the full parameter space by the following *Monte-Carlo experiment*:

- Draw k random instances of the circuit using the known probability distribution.
- If the fault efficiency of each of the k instances is equal to or larger than FE_{min} , assume that Eq. 2 holds.

However, this conclusion is not universal but rather holds with a certain confidence γ that is related to the number of repetitions k and the probability c . For example, if $k = 1$, only one random instance is evaluated and if its fault efficiency exceeds FE_{min} , the test set fulfills Eq. 2. There is still a probability that this particular random instance simply happened to have an excessively high FE by chance and that the performance of the test set T on other instances is worse than FE_{min} . Therefore, the confidence γ for guaranteeing the probability of achieving FE_{min} is considerably low. On the other hand, for $k = 100$ the test set must exceed FE_{min} on 100 random instances in a row which raises the confidence. We will now establish the formal relationship between k , c and γ .

Assume that the *actual* probability for test set T having a fault efficiency of FE_{min} with respect to N random instances is \tilde{c} . In other words, this means that out of N instances, FE of T will exceed FE_{min} for $(\tilde{c} \cdot N)$ instances. Running the Monte-Carlo experiment as described above will lead to a positive result and find a sequence of k instances where the FE of each instance is larger than or equal to FE_{min} with a probability of \tilde{c}^k . Now, the confidence level γ captures the probability that this positive result of the Monte-Carlo experiment implies Eq. 2. This is best formulated using the opposite event, namely that the outcome of the Monte-Carlo experiment is positive but the actual probability \tilde{c} is less than the desired probability c . Hence, the result of Monte-Carlo experiment was positive and misled to the belief that the probability is no less than c while in reality it was $\tilde{c} < c$.

The probability for this opposite event is $(1 - \gamma)$. Therefore, the probability of the positive result of the Monte-Carlo experiment, or \tilde{c}^k , must be less than $(1 - \gamma)$ for all $\tilde{c} < c$. From $\tilde{c}^k \leq 1 - \gamma$ for all $\tilde{c} < c$, it follows $c^k \leq 1 - \gamma$. This implies that in order to obtain a desired confidence level γ , the following number k of instances must be considered:

$$k \geq \left\lceil \frac{\ln(1 - \gamma)}{\ln(c)} \right\rceil. \quad (3)$$

Eq. 3 delivers k with the following property. If a test set T is applied to k randomly generated instances and exceeds fault efficiency FE_{min} for all of them, then the probability that an arbitrary randomly generated circuit instance will exceed fault efficiency FE_{min} is c with confidence γ . This property will be utilized in the adaptive test generation procedure in Section III.

III. CONFIDENCE-GUIDED TEST GENERATION

The confidence-guided test generation procedure aims at producing a test set T that achieves a certain desired fault efficiency FE_{min} on *any* arbitrary instance with probability c and confidence γ . As explained in Section II-C, a test set T fulfills this condition if it exceeds FE_{min} for k randomly generated circuit instances, where k is calculated using Eq. 3.

The pseudo-code of the procedure is found in Algorithm 1. Before the algorithm is explained, key sub-routines called by the method and their relevant properties are outlined.

- **PHAETON** is a SAT-based path-oriented small-delay test pattern generator [13, 14]. Given a set of faults F and a circuit instance i , it searches, for each fault $f \in F$, for the longest path through the fault location that can be sensitized in i and calculates the test pair that does the sensitization. In this work, PHAETON uses robust path

sensitization that is not prone to invalidations but may miss detectable faults.

- *WaveSAT* is a small-delay ATPG engine that generates timed sequences of rising and falling transitions (waveforms) on the circuit lines [5] that lead to the detection of the fault. Since *WaveSAT* does not necessarily sensitize the longest possible path, its generated test pair do not tend to detect many faults. However, *WaveSAT* allows accurate classification: if no test pair is found, it is guaranteed that none exists, i. e., the fault is undetectable under the given timing assumptions. *WaveSAT* is used for two purposes: close the gaps in fault coverage by pinpointedly generating test pairs for faults missed by *PHAETON* test pair sets, and prove that a fault is undetectable in a specific circuit instance and can be excluded from fault-efficiency calculation.
- *fsim* is a delay fault simulator that runs in parallel on general-purpose graphic processing units (GPGPUs) [15]. The high degree of parallelization is beneficial when grading a large number of circuit instances that have the same structure but different delays.

To construct T with the desired property, the nominal instance i_0 and a number of further instances i_1, i_2, \dots , called the *training set*, are considered (the number of instances in the training set is not known ahead of time, as new instances are added to the training set until the test set reaches the required quality). The initial T is generated to obtain 100% fault efficiency on the nominal instance i_0 in Lines 1 through 4 of Algorithm 1. *PHAETON* is used to quickly generate a good test set and *WaveSAT* to cover the missed faults identified by *fsim*. In Line 5, the number k of circuit instances that need to be considered to obtain the fault-efficiency estimate with the desired confidence γ is calculated according to Eq. 3. Then, Lines 7 through 16 are repeated until the test set T fulfills the FE target, that is, exceeds FE_{\min} for k consecutive random instances i . This is implemented as follows.

The random instance i is generated in Line 8, and the faults F_{det} detected by T are determined by fault simulation in Line 9. The set of undetectable faults F_{nd} is obtained by running *WaveSAT* and collecting all faults shown to be undetectable. In addition, *WaveSAT* generates *top-up test pairs* for faults not detected by the initial T ; these pairs are collected in set T_{tu} . If at least one of k instances does not reach fault efficiency of FE_{\min} , the top-up pairs are added to T and another k instances are considered (this is achieved by setting the loop variable j to 0). The iterations stop when T exceeds FE_{\min} for k instances in a row. This means that T is of sufficient quality and fulfills the specification; therefore, this T is returned in Line 18.

IV. EXPERIMENTAL RESULTS

A. Explicit test generation for random circuit instances

To evaluate the need for considering variations during test generation, we performed the following experiment for combinational cores of industrial circuits provided by NXP.

Algorithm 1 Confidence-guided test generation algorithm.

Procedure **conf_testgen**($C, F, FE_{\min}, c, \gamma$)

Inputs: Circuit C with random parameter distribution, fault set F , fault efficiency target FE_{\min} , probability c , confidence γ .

Output: Test set T with fault efficiency $\geq FE_{\min}$ with probability c and confidence γ .

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(1)  $i := \text{nominal\_instance}(C)$ ;
(2)  $T := \text{PHAETON}(i, F)$ ;
(3)  $F_{\text{det}} := \text{fsim}(i, T, F)$ ;
(4)  $T := T \cup \text{WaveSAT}(i, F \setminus F_{\text{det}})$ ;
    //  $T$  has 100% fault efficiency on the nominal instance
(5)  $k := \left\lceil \frac{\ln(1-\gamma)}{\ln(c)} \right\rceil$ ;
(6)  $T_{\text{tu}} := \emptyset$ ;  $j := 1$ ;
(7) while ( $j \leq k$ ) begin
(8)    $i := \text{random\_instance}(C)$ ;
(9)    $F_{\text{det}} := \text{fsim}(i, T, F)$ ;
(10)   $\tilde{T} := \text{WaveSAT}(i, F \setminus F_{\text{det}})$ ;  $T_{\text{tu}} := T_{\text{tu}} \cup \tilde{T}$ ;
(11)   $F_{\text{nd}} := F \setminus \text{fsim}(i, T, F \setminus F_{\text{det}})$ ;
(12)  if  $\left( \frac{|F_{\text{det}}|}{|F \setminus F_{\text{nd}}|} < FE_{\min} \right)$  then begin
    // Insufficient  $F$  for one of  $k$  instances, enlarge  $T$ 
(13)     $T := T \cup T_{\text{tu}}$ ;  $T_{\text{tu}} := \emptyset$ ;
(14)     $j := 0$ ; // Evaluate  $k$  new instances
(15)  end if
(16)   $j := j + 1$ ;
(17) end while // Sufficient  $FE$  for  $k$  instances in a row,  $T$  finished
(18) return  $T$ ;
end conf_testgen

```

For each circuit, we generated one nominal instance i_0 with all delays set to their nominal values, and 1000 instances i_1, \dots, i_{1000} with delays described by independent random variables according to a Gaussian distribution with standard deviation of 20%. We considered small-delay faults at 100 randomly chosen but fixed fault locations (gates) with nine different sizes equiprobably distributed across the clock cycle, i.e., 10% of t_{obs} , 20% of t_{obs} , through 90% of t_{obs} . A substantial portion of these faults are provably undetectable, as no sensitizable path of sufficient length exists through their location. We generated a number of test sets T_N with complete coverage of detectable faults (fault efficiency of 100%) on the instances i_0 through i_N . This has been done by running *PHAETON* on i_0 , simulating the generated test set on instances i_0, \dots, i_N , and using *WaveSAT* to either detect all undetected faults or classify them as provably undetectable. For example, test set T_{10} detects all detectable faults in the nominal instance i_0 and ten further instances i_1 through i_{10} .

The quality of test sets T_N is reported in Columns 6–15 of Table I (the numbers for 5-detect and 10-detect timing-aware transition fault test sets generated by a commercial tool by sensitizing the longest path through a fault location are given in Columns 4 and 5 for reference). Column 3 contains the number of detectable faults aggregated over all 1001 instances. For each test set, the size of the test set $|T|$, the number $|F|$ of detected faults in all instances and the fault efficiency FE , obtained by dividing this number by the value in Column 3, are quoted. It can be seen that the fault efficiency of variation-

TABLE I
TEST SET SIZE $|T|$, NUMBERS OF DETECTED FAULTS $|F|$ AND FAULT EFFICIENCY FE [%] OVER 1001 CIRCUIT INSTANCES (9 FAULT SIZES).

Circuit	Gates	Detectable Faults	Timing-aware for i_0			Variation-aware test sets T_N generated for instances i_0, \dots, i_N										
			5-detect	10-detect		T_0	T_5	T_{10}	T_{25}	T_{50}	T_{100}	T_{200}	T_{400}	T_{500}	T_{800}	T_{1000}
p45k	25679	210290	$ T $	110	215	317	363	386	421	449	489	531	581	602	655	686
			$ F $	176466	181531	196490	204601	207241	208611	209250	209679	210016	210147	210185	210259	210290
			FE	83.916	86.324	93.438	97.295	98.550	99.202	99.505	99.709	99.870	99.932	99.950	99.985	100.000
p78k	70475	434292	$ T $	41	75	605	672	734	849	993	1175	1401	1675	1774	1967	2096
			$ F $	352426	364983	416520	420808	424169	427853	430093	431759	432878	433662	433859	434150	434292
			FE	81.150	84.041	95.908	96.895	97.669	98.517	99.033	99.417	99.674	99.855	99.900	99.967	100.000
p89k	58638	155318	$ T $	95	192	284	309	322	355	381	416	462	515	538	575	594
			$ F $	118412	124450	149104	151967	152642	153569	154236	154780	155027	155205	155243	155298	155318
			FE	76.238	80.126	95.999	97.842	98.277	98.874	99.303	99.654	99.813	99.927	99.952	99.987	100.000
p100k	61066	201883	$ T $	77	145	303	349	372	432	481	532	613	709	744	822	873
			$ F $	167021	172682	187173	194385	196259	199152	200180	200848	201331	201644	201716	201826	201883
			FE	82.732	85.536	92.714	96.286	97.214	98.647	99.156	99.487	99.727	99.882	99.917	99.972	100.000

aware test sets dramatically exceeds the values achieved by the conventional timing-aware test sets even if a small number of instances is considered. Moreover, the fault efficiency of test set T_N quickly saturates with growing N , exceeding 99% for low double-digit values of N . The test set size $|T_N|$ increases only moderately with the size of the training set. For $N \approx 50$ the average $|T_N|$ is about twice the size of the 10-detect test set, while reaching a fault efficiency of more than 99%.

B. Confidence-driven test generation

The application of the confidence-guided test-generation algorithm *conf_testgen* is reported in Table III. Recall that the algorithm takes the circuit, the fault set (same as in the first experiment), the fault-efficiency target FE_{\min} , the confidence level γ and the probability c as inputs, where c stands for likelihood that the generated test set T exceeds FE_{\min} on a randomly generated circuit instance. Results are reported for a number of γ values (groups of columns) and a number of FE_{\min} with c set to the same value (groups of rows). Since both γ and c affect the number of instances k that have to be considered in the Monte-Carlo experiments (Eq. 3), the respective value of k is denoted in Table II for each combination.

Table III shows six columns for each experiment performed: The size $|I_\gamma|$ of the training set, that is, the number of random circuit instances considered before test $T(I_\gamma)$ passed the statistical test of Eq. 3; the size $|T(I_\gamma)|$ of this test set; the pattern generation time t_p ; and the simulation time t_s ; the fault efficiency FE achieved by this test set on 200 new random instances (*not* included in the training set); and the percentage \tilde{c} of instances with $FE \geq FE_{\min}$ among these 200 random instances.

TABLE II
VALUE OF k DEPENDING ON TARGET PROBABILITY c AND CONFIDENCE γ .

	$\gamma = 0.95$	$\gamma = 0.98$	$\gamma = 0.99$
$c = 0.98$	k=149	k=194	k=228
$c = 0.985$	k=199	k=259	k=305

It can be seen that the obtained test sets are of extremely high quality. The average FE substantially exceeds the target FE_{\min} in all cases. The percentage \tilde{c} of individual instances for which the test sets exceed FE_{\min} is always much larger than the target value c , and in most cases even reaches 100%. The runtimes associated with the confidence-driven test generation are moderate and do not exceed a few hours, even for large circuits under high confidence and probability targets. Recall that the test sets are evaluated on random circuit instances unrelated to the ones for which they have been generated. Interestingly, while the number of considered instances I_γ appears to scale with k , the differences between the test set sizes and fault efficiencies reached for a fixed circuit are rather small.

We also applied the test sets (that were obtained assuming uncorrelated variations) to a population that consists of 1025 correlated instances. The correlated instances were drawn from a two-dimensional parameter space that describes gate delay distributions with multiple gradients and different orientations in order to model spatial correlations. This parameter space has the same mean and variance over all instances as compared to the random cases.

The results can be found in Table IV. Columns FE quote the average fault efficiency of the generated test set among all the correlated instances, and columns \tilde{c} contain the percentage of instances with FE exceeding FE_{\min} . It can be seen that test sets generated assuming no correlations are highly effective even in presence of (unexpected) correlations. One explanation for this is the fact that the set of possible combinations of delay value in the uncorrelated case is a strict superset of any corresponding set for the correlated case. In other words, any delay combination that can occur in presence of correlations can also occur in absence of correlations, and there is some chance that a test pair for this combination is included in test set T .

All experiments have been conducted on a host system equipped with Intel Xeon processors clocked at 2.8GHz, 256GB RAM and NVIDIA GeForce GTX 480 graphics processing cards with 1.6GB RAM.

TABLE III

NUMBER OF CONSIDERED INSTANCES $|I_\gamma|$, TEST SET SIZE $|T(I_\gamma)|$, PATTERN GENERATION TIME t_p , SIMULATION TIME t_s , FAULT EFFICIENCY FE AND ACTUAL PERCENTAGE OF SUFFICIENTLY COVERED INSTANCES \tilde{c} FOR FAULT-EFFICIENCY TARGET FE_{\min} , TARGET PROBABILITY c AND CONFIDENCE γ .

Fault Efficiency Target FE_{\min} (set equal to target probability c)	Circuit	$\gamma = 0.95$						$\gamma = 0.98$						$\gamma = 0.99$						
		$ I_\gamma $	$ T(I_\gamma) $	t_p	t_s	FE	\tilde{c}	$ I_\gamma $	$ T(I_\gamma) $	t_p	t_s	FE	\tilde{c}	$ I_\gamma $	$ T(I_\gamma) $	t_p	t_s	FE	\tilde{c}	
		[#]	[#]	[s]	[s]	[%]	[%]	[#]	[#]	[s]	[s]	[%]	[%]	[#]	[#]	[s]	[s]	[%]	[%]	
0.98	p45k	248	547	1109	808	99.679	100.0	293	547	1109	957	99.686	100.0	327	547	1109	1069	99.693	100.0	
	p78k	287	1351	532	2214	99.495	100.0	332	1351	532	2614	99.497	100.0	366	1351	532	2916	99.492	100.0	
	p89k	248	458	1322	1519	99.617	100.0	293	458	1322	1827	99.626	100.0	327	458	1322	2062	99.658	100.0	
	p100k	316	631	6189	1930	99.625	99.5	361	631	6189	2174	99.645	99.5	604	789	6587	3811	99.812	99.5	
	0.985	p45k	301	517	1108	983	99.681	99.5	581	656	1110	1924	99.893	100.0	627	656	1110	2078	99.883	100.0
		p78k	352	1337	532	2606	99.513	99.5	646	1849	535	5861	99.794	100.0	692	1849	535	6387	99.768	100.0
		p89k	314	430	1321	2050	99.648	99.0	621	565	1323	4225	99.855	100.0	667	565	1323	4551	99.875	100.0
		p100k	575	755	6532	3801	99.792	99.0	635	755	6532	4194	99.794	99.0	681	755	6532	4497	99.784	99.0

TABLE IV

TEST SET VALIDATION ON 1025 CORRELATED INSTANCES.

Fault Efficiency Target FE_{\min} (set equal to target probability c)	Circuit	$\gamma = 0.95$		$\gamma = 0.98$		$\gamma = 0.99$	
		FE	\tilde{c}	FE	\tilde{c}	FE	\tilde{c}
		[%]	[%]	[%]	[%]	[%]	[%]
0.98	p45k	99.276	91.4	99.276	91.4	99.276	91.4
	p78k	99.825	99.9	99.825	99.9	99.825	99.9
	p89k	99.308	85.2	99.308	85.2	99.308	85.2
	p100k	99.342	96.6	99.342	96.6	99.668	99.5
0.985	p45k	99.257	81.9	99.674	92.1	99.674	92.1
	p78k	99.835	99.9	99.888	99.9	99.888	99.9
	p89k	99.371	78.1	99.599	88.2	99.599	88.2
	p100k	99.668	97.6	99.668	97.6	99.668	97.6

V. CONCLUSIONS

We presented the first test generation method that produces test sets with proven statistical performance under variations. Its core procedure iteratively enriches the test set by explicitly considering random circuit instances, guided by an estimated confidence level. The method employs the latest SAT-based timing-aware test generation engines which can provide complete testability characterization of all faults in a circuit instance, thus extending the concept of fault efficiency to variation-aware testing for the first time. The obtained test sets guarantee a minimum fault efficiency on all instances with user-defined probability and confidence bounds. The method is applicable to industrial circuits.

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